

ELIMINATION OF LOW ORDER HARMONICS IN MULTILEVEL INVERTERS USING GENETIC ALGORITHM

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ABSTRACT –

The selective harmonic elimination pulse width modulation (SHEPWM) switching strategy has been applied to multilevel inverters to remove low order harmonics. Naturally, the related equations do not have feasible solutions for some operating points associated with the modulation index (M). However, with these infeasible points, minimizing instead of eliminating harmonics is performed. Thus, harmful harmonics such as the 5th harmonic still remains in the output waveform. Therefore, it is proposed in this paper to ignore solving the equation associated with the highest order harmonics. A reduction in the eliminated harmonics results in an increase in the degrees of freedom. As a result, the lower order harmonics are eliminated in more operating points.

A 9-level inverter is chosen as a case study. The genetic algorithm (GA) for optimization purposes is used. Simulation results verify the proposed method.

KEYWORDS - Genetic Algorithm (GA), Multi-level Inverter, Selective Harmonic Elimination Pulse Width Modulation (SHEPWM)

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I. INTRODUCTION

In recent years, electrical industries have expanded and the variety of loads has increasingly grown. Recently, the industry has begun to apply high-voltage high-power equipment that has reached the megawatt range. Today, the direct connection of a single semiconductor switch to a system with medium sized voltage grids (2.3, 3.2, 4.16 and 6.9 KV) is too difficult [1]. To overcome the limitations on semiconductor voltage and current ratings, some type of series connection will be necessary. Therefore, multilevel inverters have been introduced. Due to their ability to synthesize waveforms with a better harmonic spectrum and attain a higher voltage without transformers, they have been receiving increasing attention in especially in distributed generation systems (DG) [2], high voltage direct current transmissions (HVDC) [3], electrical drives [4] and flexible AC transmission systems (FACTS) devices [5]. Today, there are many switching strategies which are applied to multilevel inverter topologies. Some of these switching strategies are mentioned in [6]. Selective harmonic elimination pulse width modulation (SHEPWM) [6]–[8] is the most famous switching strategy. The aim of this method is the eliminating low order harmonics, if it is possible, or at least to minimize them [9]. The nature of these equations is nonlinear. Some iterative methods such as: the Newton- Raphson (N-R) have been used to solve the equations [10].

Iterative methods depend mainly on the initial guess and divergence problems are likely to occur. Resultant theory as an analytical method has been proposed in [11] to find all feasible solutions. However, this method is not applicable for a high number of levels. genetic algorithm (GA) has been proposed in [12]. The GA is simple and applicable to problems with any number of levels, without the extensive derivation of analytical expressions, for both eliminating and minimizing harmonics.

This algorithm is chosen for the optimization goals. In conventional SHEPWM, two sets of modulation indexes (M) are specified. In the first set, the equations have feasible solutions, whereas in the second set there are not any analytical solutions. Therefore, when using these optimizing methods, the low order harmonics are minimized as much as possible but not removed completely. As a result, low order harmonics such as: the 5th harmonic still remain in the output voltage. There are S switching angles associated with S cells. These angles are considered as variables or degrees of freedom. Thus, S equations can be satisfied. $S-1$ equations

are devoted to eliminating the (S-1) low order harmonics starting from the 5th order for 3-phase applications or from the 3rd order for single phase applications. Also, one equation is dedicated to satisfying the fundamental component. The modulation index (M) is defined as a symbol of the fundamental component. In some Ms, which are focused on here, there are no feasible solutions, whereas some loads need to work in whole range of Ms [13].

In this paper, a method to categorize all values of M in S groups is proposed. For each group, special equations are solved. In the first group, all S equations are satisfied and S-1 harmonics are perfectly removed. In the second case,

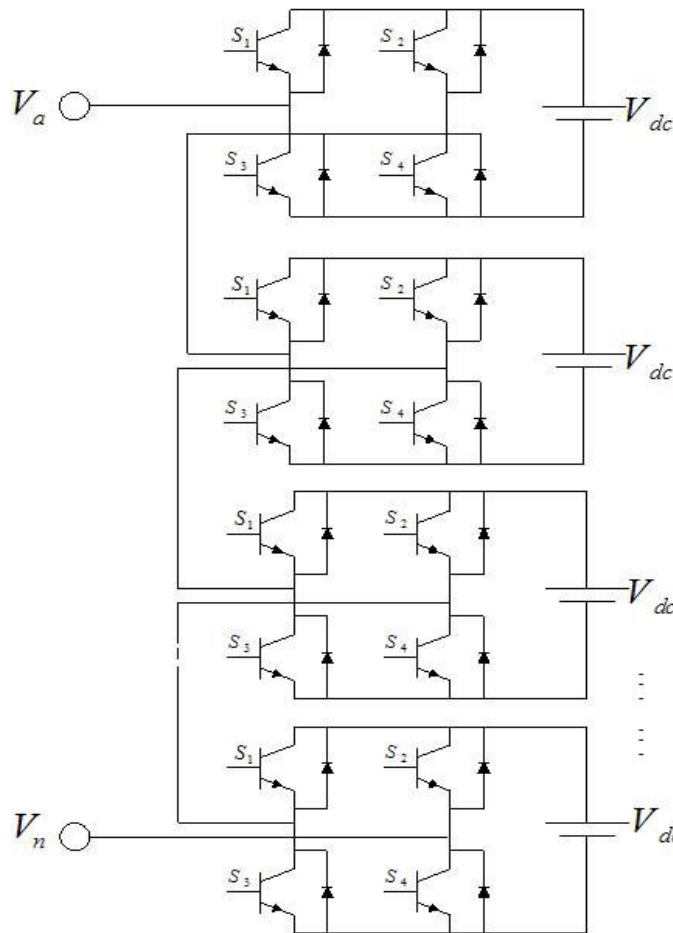


Fig. 1. Cascaded multilevel inverter.

the elimination of one equation is ignored. Thus, the new equations have more feasible solutions when compared with the first group. Those Ms which are not dependent on the first group are devoted to the second group. Subsequently, S-2 equations are solved and a third group is constructed. Finally, the last group is dedicated to satisfying only the fundamental component.

These assumptions result in harmful lower order harmonics, such as 5th harmonic, being removed for more intervals of M_s . Simulation based results confirm that the purposed method provides a suitable waveform with less harmonic pollution. A 9-level inverter is chosen as a case study.

II. STRUCTURES OF MULTI-LEVEL INVERTERS

Three main topologies have been proposed for multilevel inverters. These topologies are diode clamped multi-level inverters (DCMI), flying-capacitor multi-level inverters (FCMI) and cascaded multi-level inverters with separate DC sources [14].

The cascaded multi-level inverter shown Fig. 1 has many advantages such as circuit layout flexibility, no extra clamping diodes or voltage balancing capacitors, an easily adjusted number of output voltage levels by adding or removing fullbridge cells, and the lowest number of components when compared with other inverters [15]. In addition, the switching devices turn on and turn off only one time per cycle to overcome the switching loss problem [16]. This multi-level inverter is made from several full-bridge inverters. The AC outputs for each different level of the fullbridge inverters are connected in series so that the synthesized voltage waveform becomes the sum of the inverter outputs. Fig. 2 shows the output waveform of a 9-level inverter. The switching angles of the multilevel inverter which are considered as variables and should be determined.

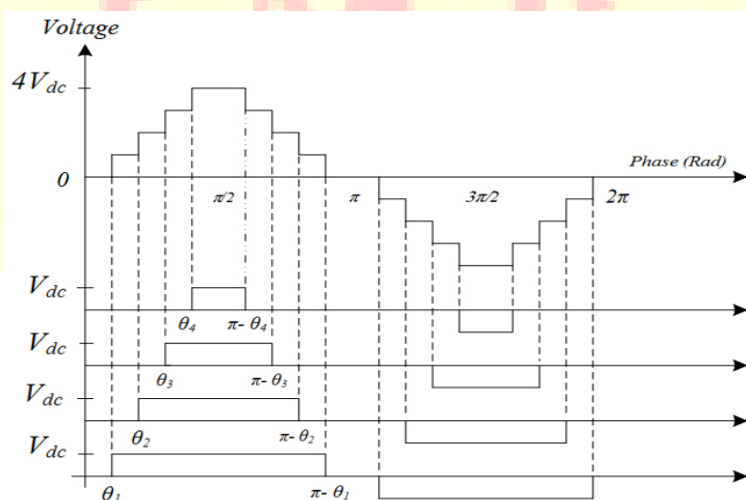


Fig. 2. Nine level waveform of cascaded multilevel inverter

V_{dc} is the voltage of the DC sources which are shown in Fig. 1. Here, it is supposed that the voltages of all of the DC sources are equal. Each full-bridge inverter produces a three level waveform +V_{dc}; -V_{dc} and 0, and each angle θ_i is related to inverter i , where, $i = 1; 2; \dots; S$. S is the number of DC sources that is equal to number of switching angles. The number of levels (L) is calculated by $L = 2S + 1$.

III. SWITCHING STRATEGIES

Nowadays, there are several switching strategies which are applied to multilevel inverter topologies such as: 1) sinusoidal pulse width modulation (SPWM); 2) selective harmonic elimination pulse width modulation (SHEPWM); 3) minimization of total harmonic distortion (MTHD). In MTHD, without any emphasis on special harmonics, all harmonics in the same weight (i.e. THD) are minimized [17], [18]. In SHEPWM, the low order harmonics are eliminated. There are some advantages and disadvantages to these three techniques which are reported in the literature [19]. In this paper, the SHEPWM technique is developed.

A. Basic SHEPWM

In this paper, a 9-level inverter is chosen as a case study. Thus, with four angles as degrees of freedom, it is possible to satisfy the fundamental component and to eliminate three low order harmonics i.e. 5th, 7th and 11th. It is not necessary to delete the triplen harmonics, because they will be eliminated in the line to line output voltage in the Y connection, automatically. In other words, for a 9-level inverter, the following equation should be solved:

$$\begin{aligned}
 M &= [\cos(\theta_1) + \cos(\theta_2) + \dots + \cos(\theta_4)]/4 \\
 0 &= \cos(5\theta_1) + \cos(5\theta_2) + \dots + \cos(5\theta_4) \quad (1) \\
 0 &= \cos(7\theta_1) + \cos(7\theta_2) + \dots + \cos(7\theta_4) \\
 0 &= \cos(11\theta_1) + \cos(11\theta_2) + \dots + \cos(11\theta_4)
 \end{aligned}$$

It is necessary to determine four switching angles, namely θ_1 , θ_2 , θ_3 and θ_4 such that the equation sets (1) are satisfied. These equations are nonlinear. The whole interval of M may be categorized into three regions. In the first region, one solution set exists. In the second region, more than one solution set exists. In the last region, which this paper focuses on, there is no

solution, naturally. Using intelligent algorithms such as: a GA leads to the minimization of harmonics instead of the elimination of them.

B. Proposed Method Based on the Reduction of Eliminated Harmonics

As previously mentioned, the elimination of low order harmonics involves solving nonlinear transcendental equations. These equations are not solvable for some intervals of M (the third region). However, some loads such as electrical drives need to work in a wide range of M even in the third region. In these points, the desired harmonics cannot be completely eliminated. Instead, the minimization of harmonics leads to the presence of harmful harmonics such as the 5th order harmonic. This paper purposes to neglect from elimination the harmonics with the highest order in Eq. (1). For example, elimination of the 11th harmonic is ignored. Therefore, only the 5th and the 7th harmonics are eliminated. As a result, one degree of freedom is added to Eq. (1). Four variables are dedicated to solving the three equations. This results in a set of equations which have feasible solutions for more intervals of M . For problem generalization, if neglecting the elimination of one harmonic cannot satisfy remaining equations, it is proposed to neglect the elimination of two harmonics. Subsequently, more harmonics are ignored until the remaining equations become solvable for the whole range of the modulation index. Since, the SHEPWM equations are nonlinear, the GA is proposed to solve them.

IV. GENETIC ALGORITHM

As mentioned, SHEPWM equations are nonlinear. In order to solve these equations the genetic algorithm (GA), which is based on natural evolution and populations, is implemented. This algorithm is usually used to reach a near global optimum solution. In each iteration of the GA a new set of strings, which are called chromosomes, with improved fitness is produced using genetic operators. A more complete discussion of GAs including extensions to the general algorithm and related topics can be found in books by Davis [20], Goldberg [21], Holland [22], and Deb [23].

The structure of a simple GA consists mainly of three operators. A selection operator, a crossover operator which acts on a population of strings to perform the required reproduction and

recombination, and a mutation operator which randomly alters character values, usually with a very low probability. The effect of these random alterations is to maintain diversity within the population, thereby preventing an early convergence of the algorithm to a possibly false peak.

A. Chromosome Representation

In this study, each chromosome is taken as a possible Solution for the problem, then each chromosome is developed based on single dimensional arrays with a length of S , where S is the number of angles.

B. Initialization of the Population

For any GA it is necessary to initialize the population. The most common method is to randomly generate solutions for the entire population. All of the experiments discussed in this paper employ a completely random seeding of the initial population. Population size depends only on the nature of the problem and it must achieve a balance between the time complexity (consume for computing the fitness function and the genetic operators) and the search space measure. In this paper, the population size is set at 150.

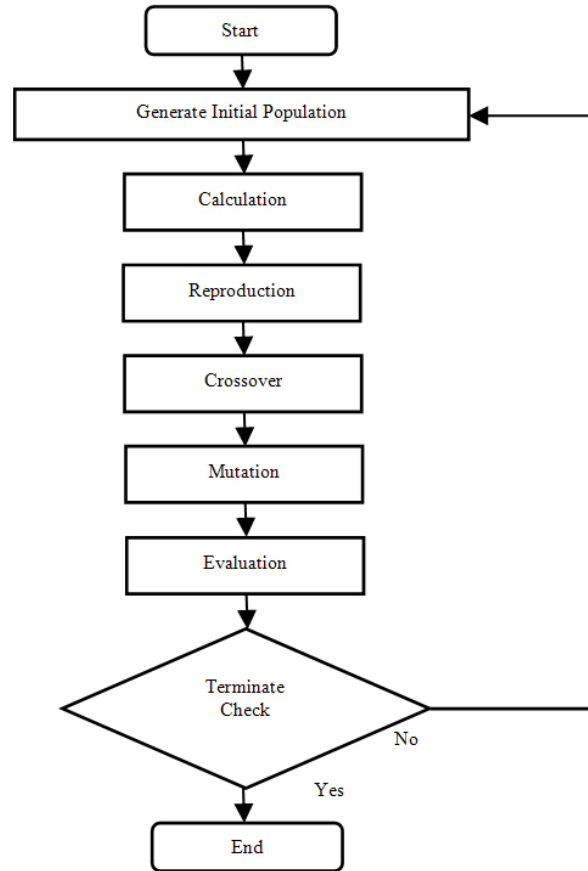


Fig. 3. Flowchart of Genetic Algorithm.

C. Reproduction

The degree of conformity of each object is calculated and an individual is reproduced under a fixed rule depending on the degree of conformity. Here, some individuals with a low degree of conformity will be screened, while individuals with a high degree of conformity will increase

D. Crossover

New individuals are generated by the method of intersection has been set up.

E. Mutation

This is performed by an operation determined by the installed mutation probability or mutation, and then a new individual is generated.

F. Evaluation of Fitness function

The fitness value is a measure of the appropriateness of a solution with respect to the original objective and the “amount of infeasibility”. The fitness function is formed by adding a penalty to the original objective function or fundamental component function [24], [25]. The GA process as well as an explanation is extracted from Refs. [26], [27].

These processes are presented in the flowchart in Fig. 3. Since finding a feasible solution is a major problem, a penalty approach is used to decrease the fitness of infeasible solutions and toward the feasible region.

V. SIMULATION RESULTS

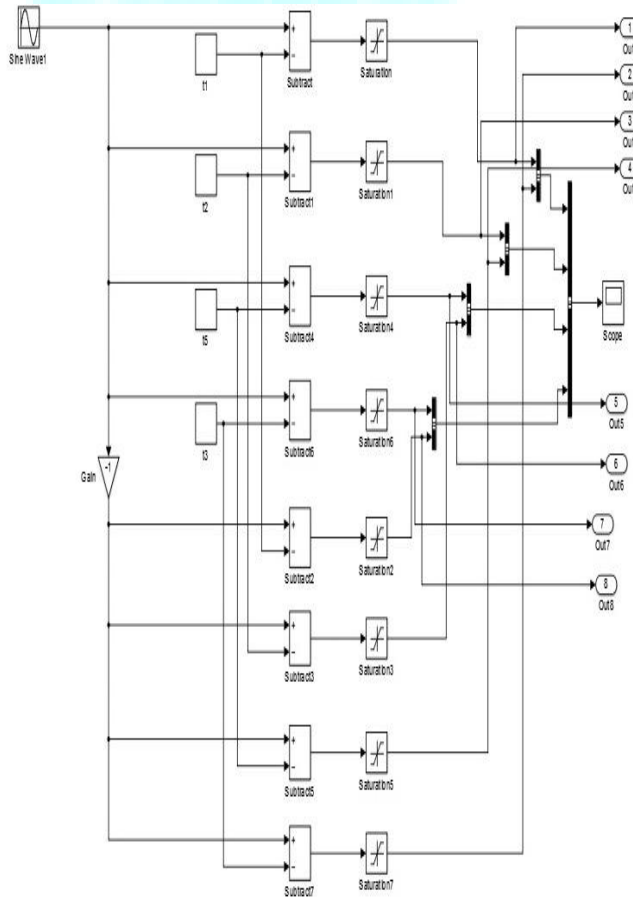


Fig 4: Nine level control strategy subsystem

A nine Level Inverter scheme been proposed to achieve the simultaneous elimination of common-mode voltage and DC-link capacitor-voltage imbalance. The operating limitations of achieving this dual task for the nine-level inverter configuration with a single DC power supply are investigated. A switching strategy for the nine-level inverter topology with two DC power supplies is proposed to achieve the dual task throughout the operating range of the drive. The proposed drive offers a simple power-bus structure with increased redundant switching states for inverter voltage vectors, and requires a lower voltage blocking capacity of power devices as compared with a conventional single five-level inverter-fed drive. As only the availability of redundant switching states for the inverter voltage vectors is exploited, the dual task is achieved without disturbing the fundamental component of the inverter output voltage and the scheme does not need any extra control circuit hardware. The performance of proposed scheme is demonstrated by the results of simulation studies, and experimentally verified induction motor drive in the linear as well as over modulation range.

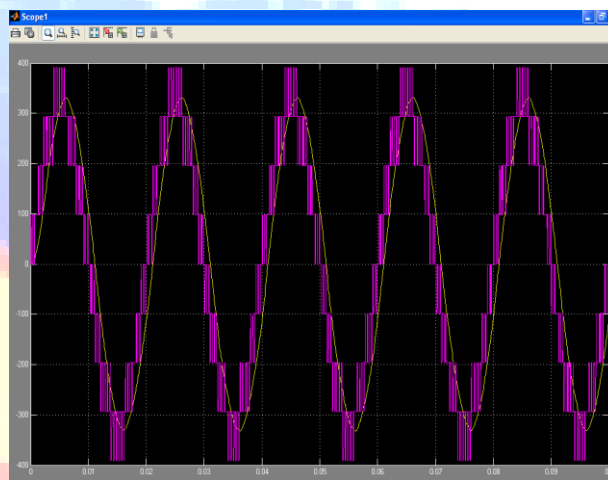


Fig 5: Output Waveform of Nine Level Inverter

For implementing the proposed technique, a code based on the genetic algorithm is written in MATLAB software. All of the results are extracted from this code.

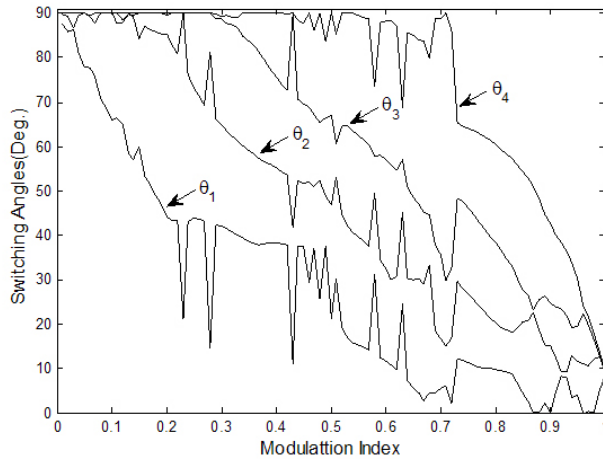


Fig. 6. Optimal switching angles versus M.

First, the GA is applied to a 9-level inverter to eliminate the 5th, 7th and 11th harmonics as a base case. The curve of the switching angles versus M is shown in Fig. (4). M varies from zero to one [0-1]. The step of increasing M is 0.01. In some Ms (e.g. [0.4-0.7]), more than one solution exists. The GA falls randomly into one of them, so jerks are observed in this figure.

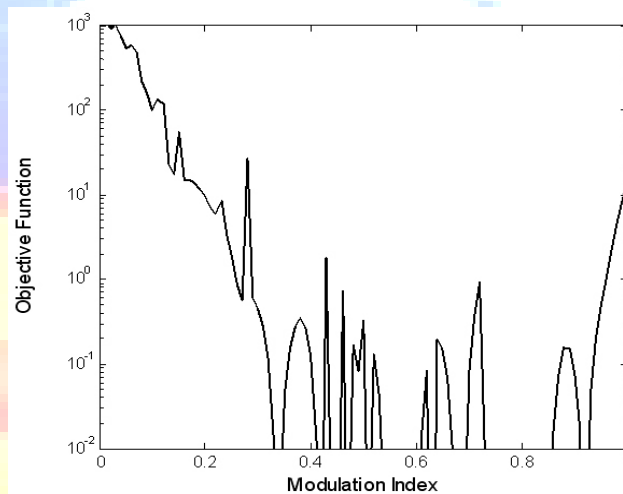


Fig. 7. Optimum values of objective function versus M.

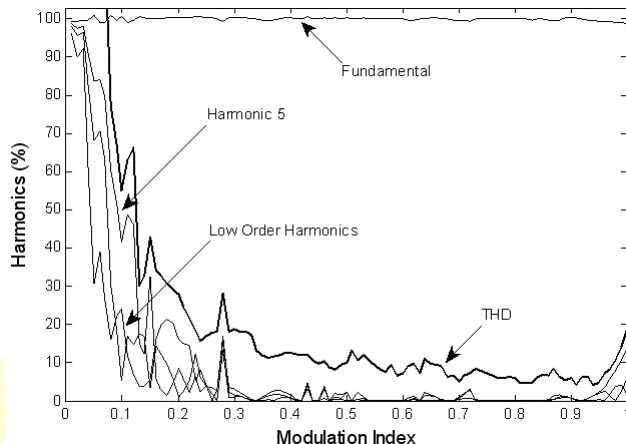


Fig. 8. Percentage of low order harmonics for base case.

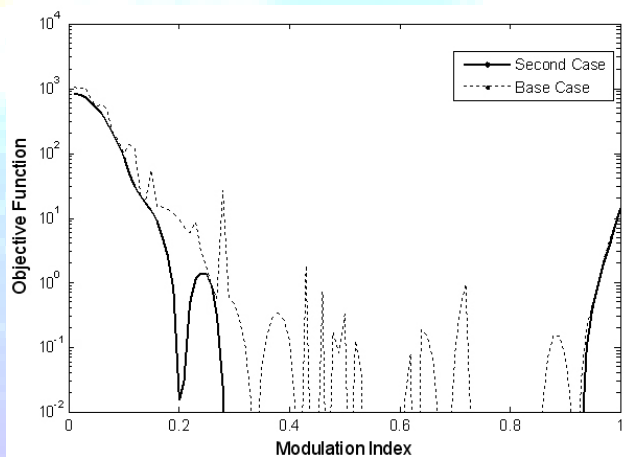


Fig. 9. Objective function for neglecting 11th Harmonic case.

Fig. (7) shows the values of low order harmonics and THD. The desired fundamental component is satisfied for all Ms. In the feasible intervals of M, low order harmonics are omitted and THD has the least values.

To provide a proper voltage with an acceptable quality of harmonics especially in unsolvable Ms, this paper purposes neglecting the elimination of the highest order harmonic.

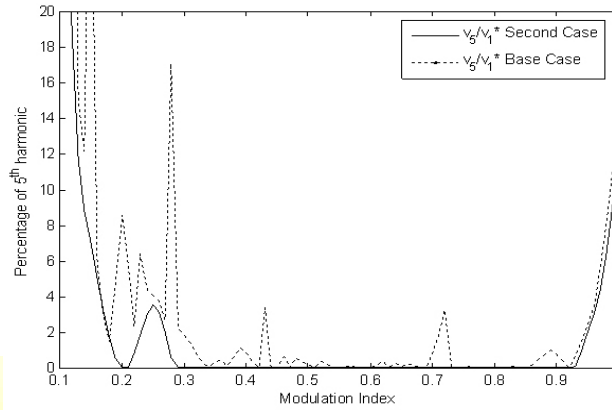


Fig. 10. Comparing 5th harmonic between two Cases.

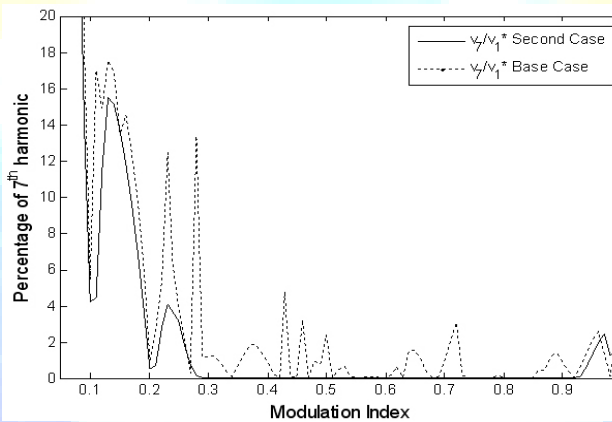


Fig. 11. Comparing 7th harmonic between two Cases.

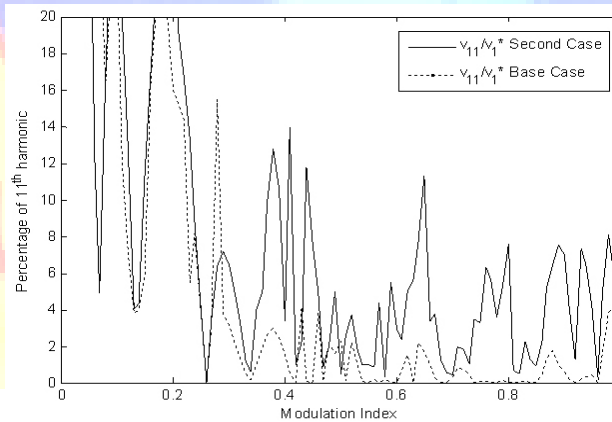
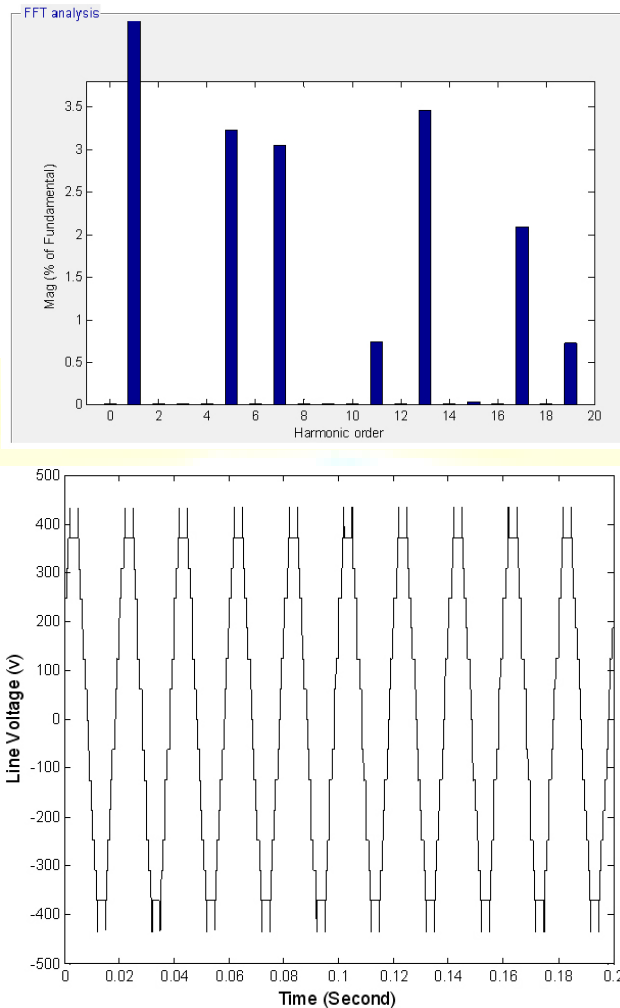


Fig. 12. Comparing 11th harmonic between two Cases.



(a)

Fig. 13. Line voltage. (a) and Frequency Spectrum of a 9-level inverter for $M=0.72$ based on base case.

V. CONCLUSIONS

In this paper, the SHEPWM switching strategy is developed to improve the quality of the low order harmonics in a wide range of modulation indexes. Normally, the equations related to SHEPWM do not have feasible solutions for all of the values of M . This leads to the presence of harmful harmonics such as: the 5th harmonic. It is suggested to reduce the number of eliminated harmonics. As a result, the 5th harmonic becomes zero for more values of M . Also, the equations are satisfied for all quantities of M .

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